



Criteria	Criteria 2- Teaching- Learning and Evaluation
Key Indicator	2.3 Teaching- Learning Process
Metric	2.3.1 Student centric methods, such as Interactive Classroom Activities methodologies used for participative learning

INTERACTIVE CLASSROOM ACTIVITIES

SMVITM faculty members engaging interactive classroom activities can significantly enhance student learning experiences and foster essential skills. The activities such as flipped classroom, Minute paper, Mind Map, quiz are used for interactive activities in the classroom. Also students are engaging themselves even outside the class room in case of flipped classroom.

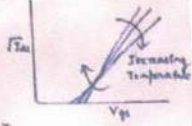
Few of the documents related to these activities are documented as a sample evidences.


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Temperature Dependence

Transistor characteristics are influenced by temperature
 μ decreases with T
 V_t decreases linearly with T
 $I_{leakage}$ increase with T
 V_t decreases linearly with T
 ON current decreases with T
 OFF current increases with T



Geometry Dependence

$L_{eff} = L_{drawn} + X_c - 2L_0$
 $W_{eff} = W_{drawn} + X_w - 2W_0$
 Layout designers draw transistors with drawn L_{drawn} & W_{drawn}
 Actual dimensions may differ from some factor X_w & X_l

Source of Leakage

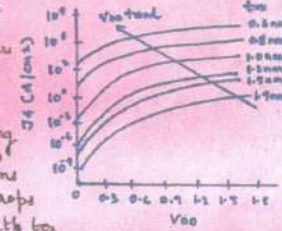
Subthreshold Conduction

In real transistors, current doesn't abruptly cut off below threshold, but rather drops off exponentially with V_{gs}
 This leakage current when transistor is normally off depends on: E_{ox} , t_{ox} , N_{bulk} , W , L , T , V_t

$$I_{sub} = I_{sub0} e^{-\frac{V_{gs} - V_{th} + V_{ds}}{nV_t}} \left(1 - e^{-\frac{V_{ds}}{V_t}}\right)$$

Gate Leakage

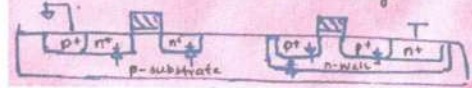
There is a finite probability that carriers will tunnel through thin gate oxide
 This results in gate leakage current flowing into gate. I_{gate} is greater for electrons
 This probability drops off exponentially with V_{gs}



$$I_{gate} = qNA \left(\frac{V_{gs}}{t_{ox}}\right)^2 e^{-B \frac{V_{gs}}{V_{ox}}}$$

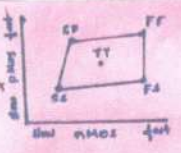
Junction Leakage

Reverse biased diodes still conduct a small amount of current that depends on:
 - Doping levels
 - Area & perimeter of diffusion
 - The diode voltage



Process Variation

Transistors have uncertainty in process parameters
 - process: L_{eff} , V_t , t_{ox} , n , μ , τ_{MOS} , τ_{PMOS}
 - Variation is around typical (1) values
 - L_{eff} : short
 - V_t : low
 - t_{ox} : thick
 - Slow (1): opposite
 - Not all parameters are independent of $nMOS$ & $pMOS$



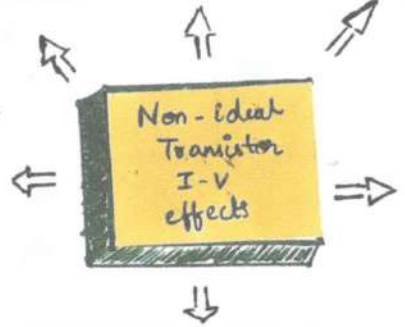
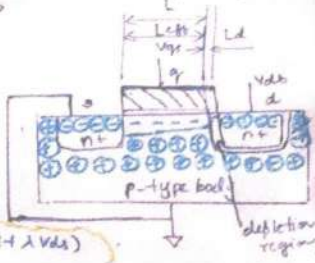
Environmental Variation

V_{DD} and Temperature also vary in time and space
 - Fast: V_{DD} : high - T: Low

Channel Length Modulation

Reverse-biased p-n junctions form a depletion region. Width of depletion L_d region grows with reverse bias V_{ds}

- $L_{eff} = L - L_d$
- Shorter L_{eff} gives more current
- I_{ds} increases with V_{ds}
- $V_{ds} \uparrow \rightarrow \text{dep} \uparrow \rightarrow \text{shortening of channel}$ i.e. $\Delta L = L_d$
- λ - empirical channel length modulation parameter
- $I_{ds} = \frac{\mu C_{ox}}{2} \frac{W}{L} (V_{gs} - V_t)^2 (1 + \lambda V_{ds})$



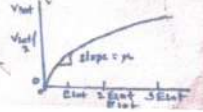
Threshold Voltage Effects

Ideal models assume V_t is constant but weakly depends on:

- Body Effect**
 - V_t varies as potential b/w source & body varies. is called body effect.
 - $V_T = V_{T0} + \gamma \left(\sqrt{\phi_s + V_{gs}} - \sqrt{\phi_s} \right)$
 - $\phi_s = \frac{2KT}{q} \ln \frac{N_{bulk}}{n_i}$; $\gamma = \frac{\sqrt{2q \epsilon_s N_{bulk}}}{E_{ox}/t_{ox}}$
 - $N_{bulk} \uparrow \rightarrow V_T \uparrow$; $C_{ox} \uparrow \rightarrow V_T \downarrow$
- Drain Induced Barrier Lowering (DIBL)**
 - Drain voltage (V_{ds}) \uparrow creates an electric field \rightarrow affects the threshold voltage.
 - As channel length \downarrow , DIBL effect shows up & variation caused in threshold voltage $V_{th} = V_t - \eta V_{ds}$
 - η = DIBL coefficient
 - High drain voltage causes current to increase
- Short channel Effect**
 - In small transistors the s/d depletion regions extend into a significant portion of the channel.
 - Impacts amount of charge required to invert channel
 - $V_t \uparrow$ with L

Electric Fields Effects

- Mobility Degradation**
 - As we increase V_{gs} , there will be more number of electrons attracted towards the channel
 - As the channel length is less & more electrons are getting accumulated in channel, mobility of electrons gets degraded because of collisions
 - This effect is termed as mobility degradation
- Velocity Saturation**
 - $v = \mu E$
 - At higher E , μ is no more constant & v varies and is due to velocity saturation effect
 - Before the velocity reaches critical value
 - $v = \frac{\mu E_{sat}}{1 + E/E_{sat}}$
 - When velocity reaches critical & greater the gain by $v = v_{sat}$

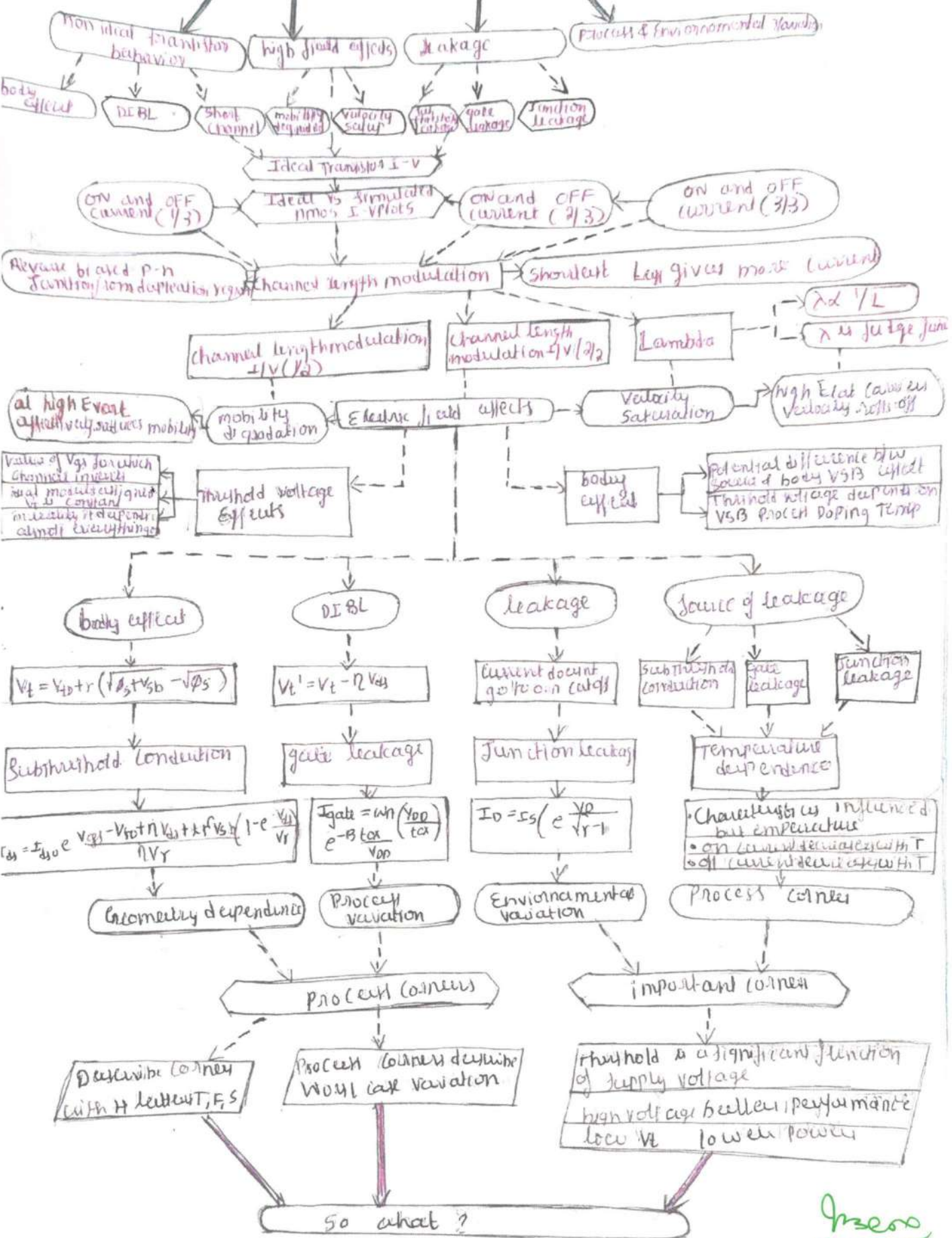


(4M20EC051) SANJANA RAO

Priscoor

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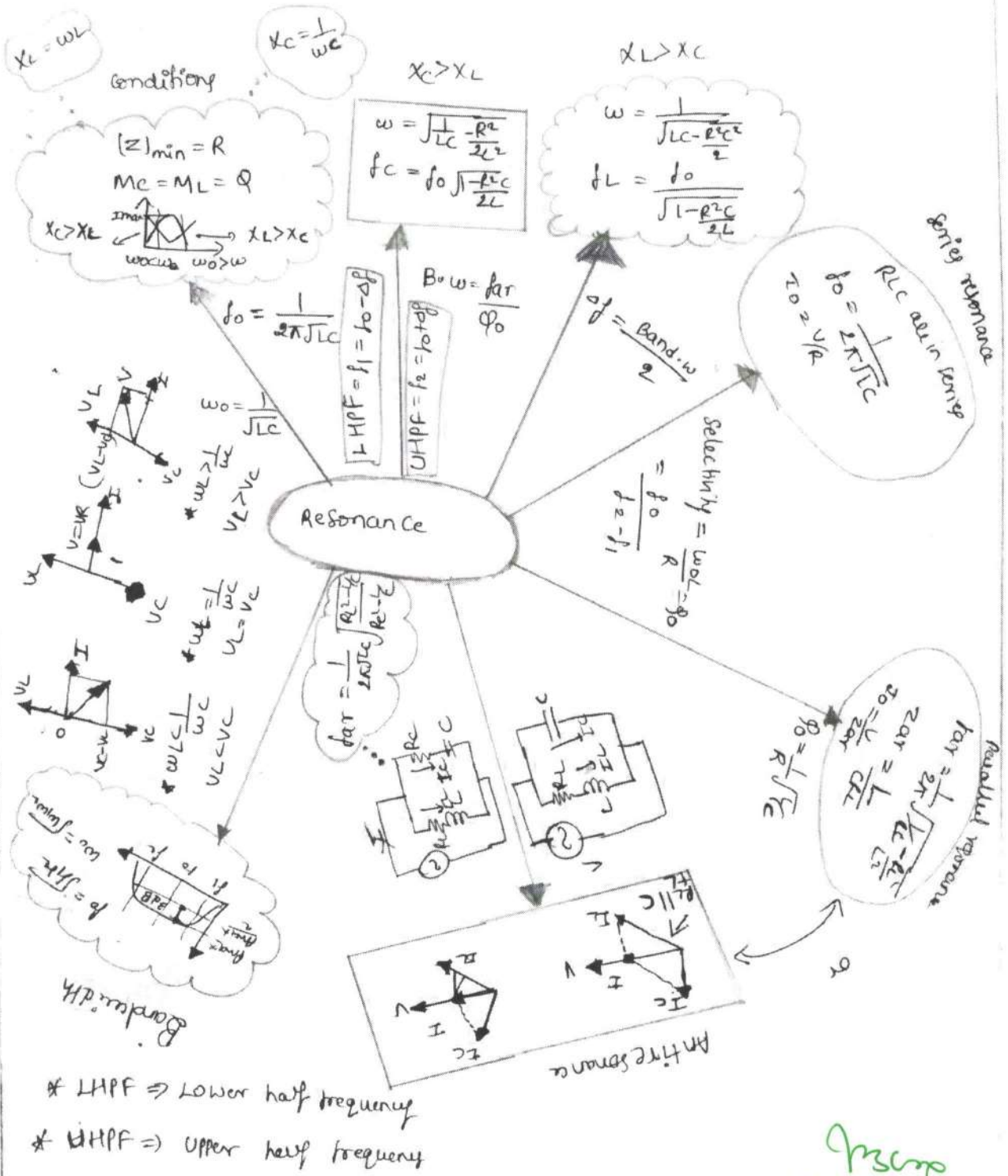
Non Ideal Transistor I-V effects



Inzero

RESONANCE :-

ANUSHA (4MW20EC012)
 MANASA (4MW20EC029)
 MEDINI (4MW20EC031)
 SONJANYA (4MW20EC033)



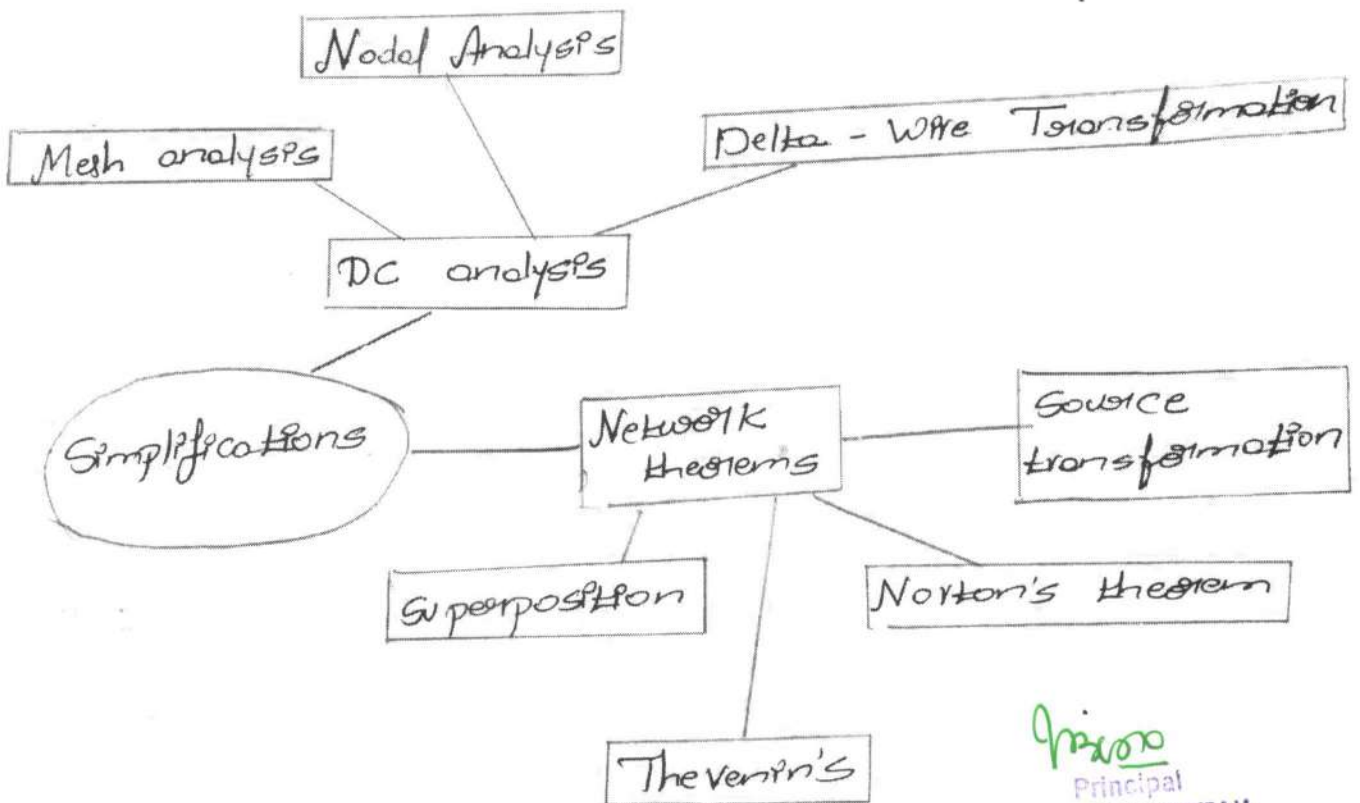
* LHPF = Lower half frequency
 * UHPF = Upper half frequency

Network Theory

Team members

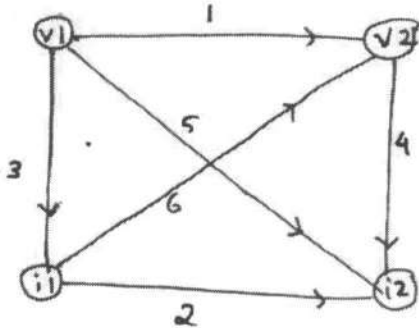
- 1] Nikhitha shetty
- 2] Manasa
- 3] Harishatha
- 4] Ankitha shet

Mind Map



Mind map for Network Parameters

Team
Nagaraj
Auston
Adarsh
Abhishek



where

- 1 → Z-Parameter
- 2 → Y-Parameter
- 3 → ABCD Parameter
- 5 → h-Parameter
- 4 → A'B'C'D' Parameter
- 6 → g-Parameter

Z-Parameter

$$V_1 = Z_{11} * I_1 + Z_{12} * I_2$$

$$V_2 = Z_{21} * I_1 + Z_{22} * I_2$$

h-Parameter

$$V_1 = h_{11} * I_1 + h_{12} * V_2$$

$$I_2 = h_{21} * I_1 + h_{22} * V_2$$

Y-Parameter

$$I_1 = Y_{11} * V_1 + Y_{12} * V_2$$

$$I_2 = Y_{21} * V_1 + Y_{22} * V_2$$

g-Parameter

$$I_1 = g_{11} * V_1 + g_{12} * I_2$$

$$V_2 = g_{21} * V_1 + g_{22} * I_2$$

ABCD Parameter

$$V_1 = A * V_2 + B * I_2$$

$$I_1 = C * V_2 + D * I_2$$

A'B'C'D' Parameter

$$V_2 = A' * V_1 - B' * I_1$$

$$I_2 = C' * V_1 - B' * I_1$$

Princip

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MINUTE PAPER

Please answer these questions using sentences

What are you hoping to learn today?

I hope to learn more about memory cell.

What did you learn today?

I learnt about Betn, rowseled and typep
of memory cell. Dynamic cell and
PLS structure.

What do you need to learn more about?

I need to learn more about PLS structure
and when it is used.



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MINUTE PAPER

Please answer these questions using sentences

What are you hoping to learn today?

We learnt about Two types of memory cell i.e Static RAM (SRAM) and Dynamic RAM (DRAM) with 1,4,6 Transistors.

What did you learn today?

Learnt about CMOS Pseudo Static RAM cell, using Transmission Gate, NMOS Pseudo Static RAM cell and their Stick diagram.

What do you need to learn more about?

We would like to learn more about the (1T1C) memory cell and its real time usage.

Arscop

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AC Quiz-1 Analysis

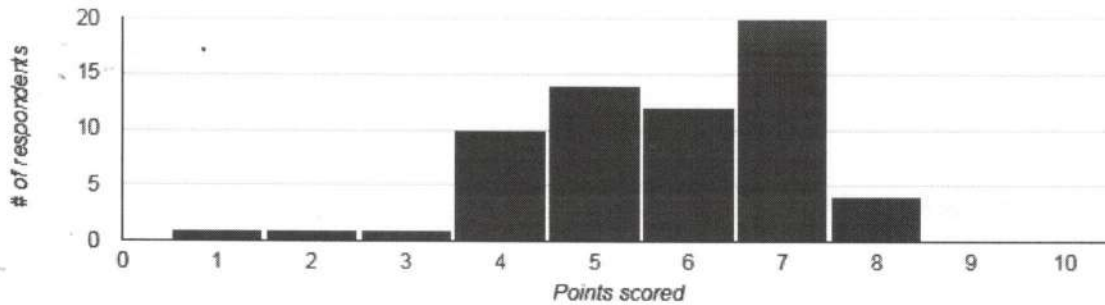
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Average
5.71 / 10 points

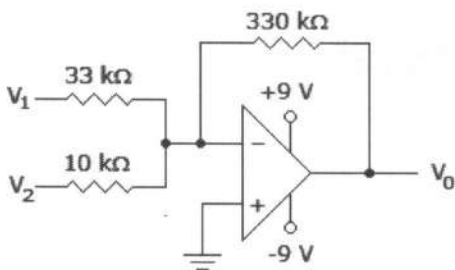
Median
6 / 10 points

Range
1 - 8 points

Total points distribution

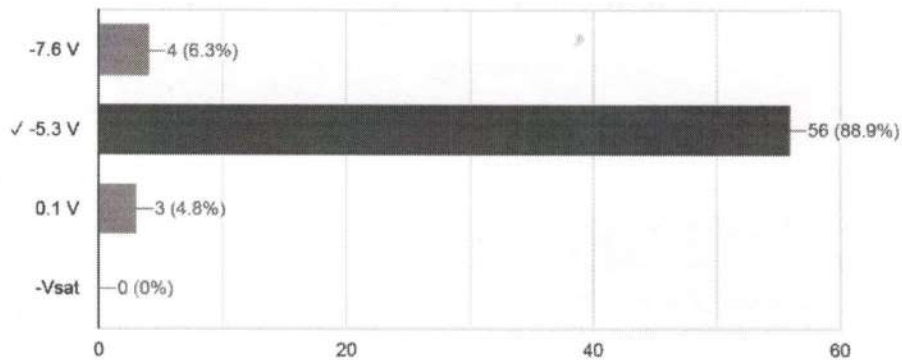


Q1



Calculate the output voltage if V₁ = 0.2 V and V₂ = 0.1 V

56 / 63 correct responses

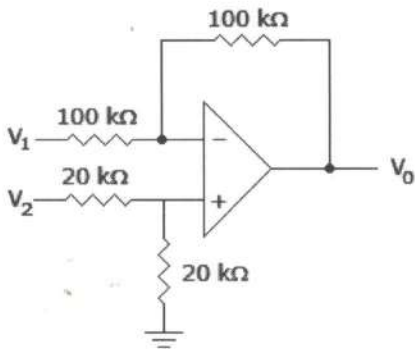


Ans

Principal

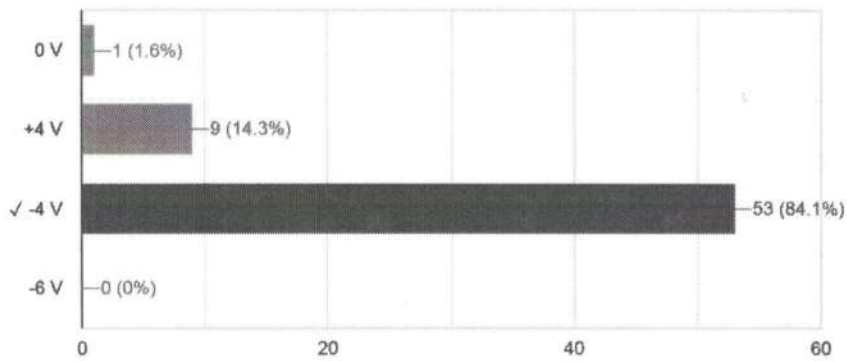
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Q2



Calculate the output voltage if V₁ = 2 V and V₂ = -2 V

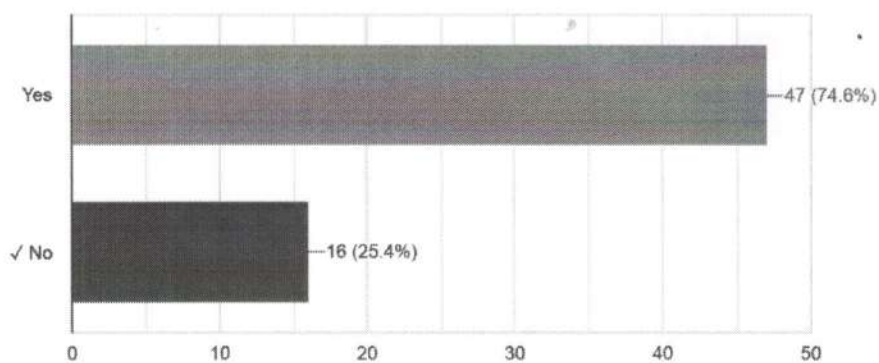
53 / 63 correct responses



Q3

Using a single op-amp in inverting configuration, can we get a voltage gain of +0.5 ?

16 / 63 correct responses



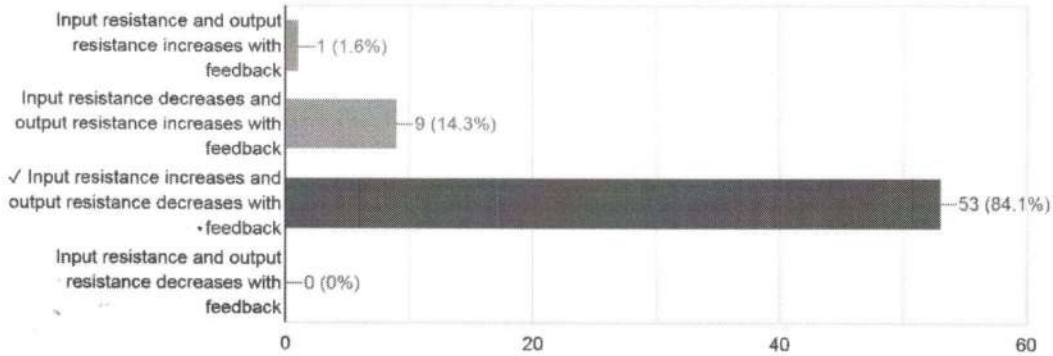
Note: In an inverting amplifier, gain is negative


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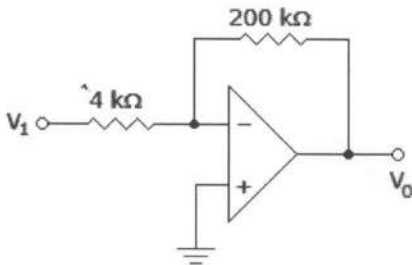
Q4

Which of the following statements is true with reference to op-amp non-inv amplifier?

53 / 63 correct responses

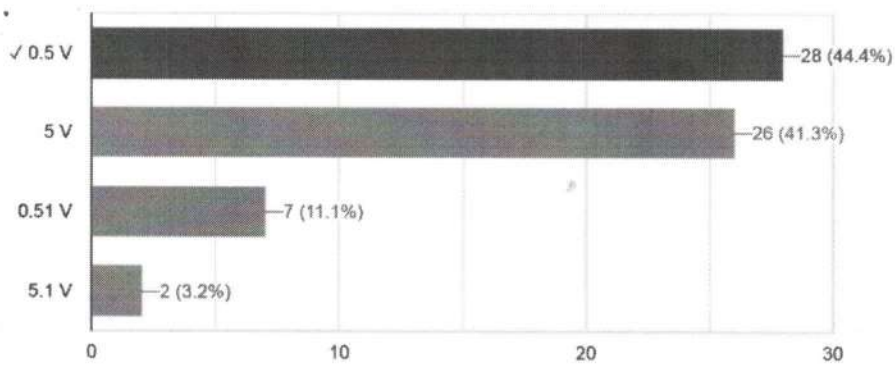


Q5



What is the amplitude of the output voltage if the input is a sinusoidal with 10 mV amplitude?

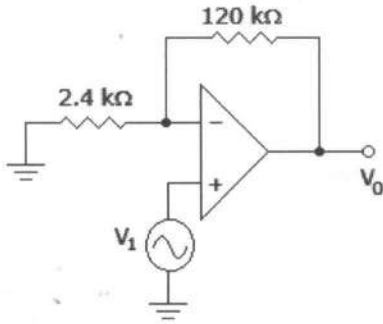
28 / 63 correct responses



Amcox
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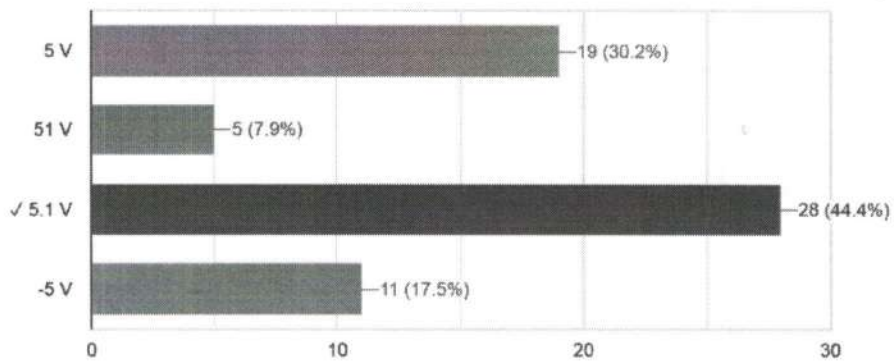
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Q6.

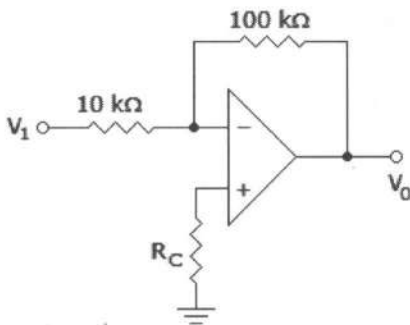


What is the amplitude of the output if the input is a sinusoidal with 100 mV amplitude?

28 / 63 correct responses

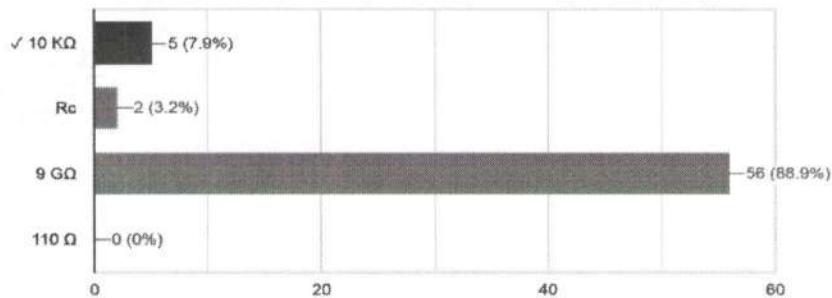


Q7.



What is the input resistance of the amplifier if $R_i = 1 \text{ M}\Omega$ and $A = 100,000$.

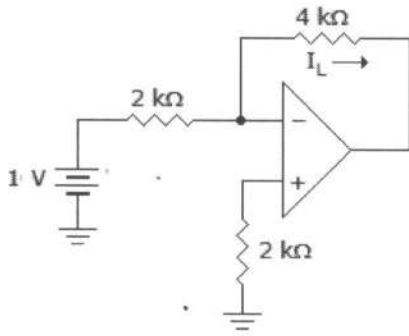
5 / 63 correct responses



Note: For inverting amplifier, $R_{if} = R_1$

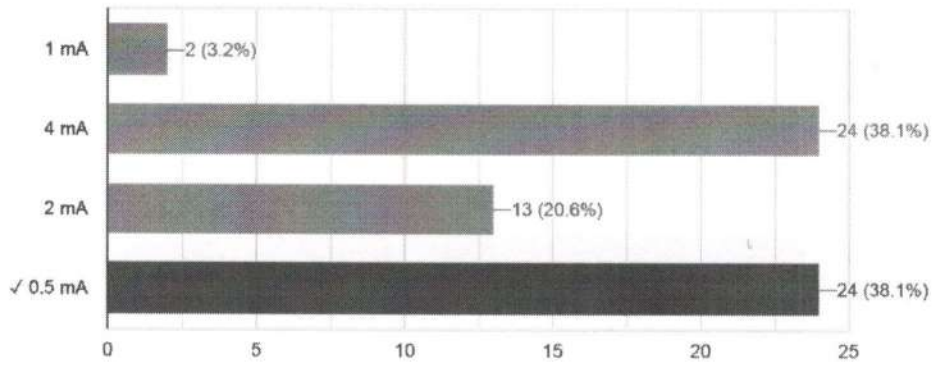

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Q8.

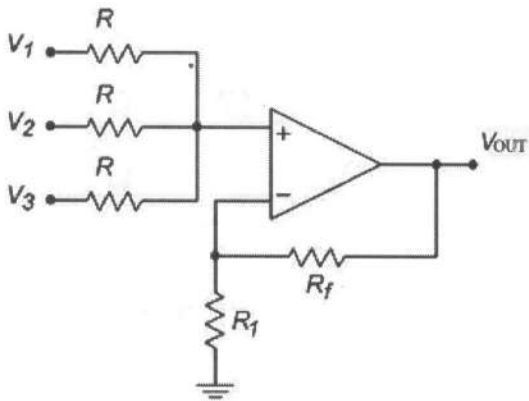


What is the value of the current I_L ?

24 / 63 correct responses

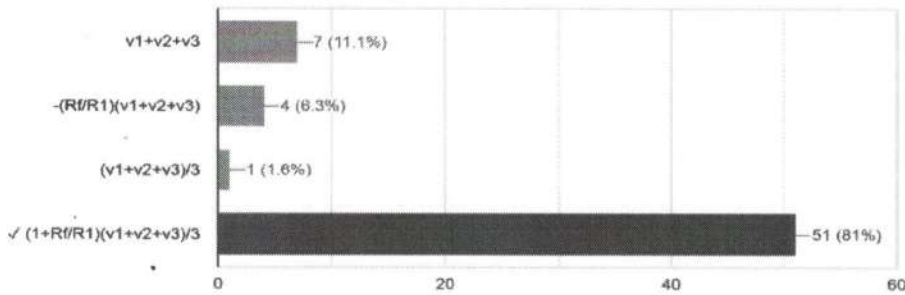


Q9.



Output voltage is

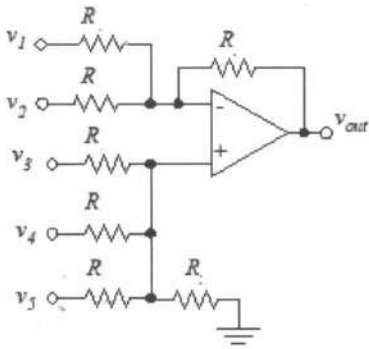
51 / 63 correct responses



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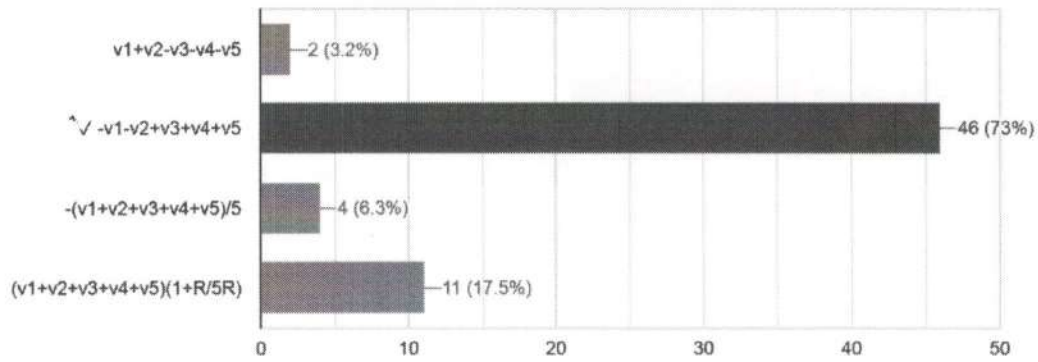
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Q10.



Output voltage is

46 / 63 correct responses




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Active Low Pass and High Pass Filter - LAB EXPERIMENT

By
Arun Upadhyaya
Asst. Professor (Sr.)
Dept. of ECE
SMVITM, Bantakal

ACTIVE LOW PASS AND HIGH PASS FILTER #ACLAB

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ASYMMETRIC ASTABLE MULTIVIBRATOR

Asymetric Frequency = 1 kHz and Duty Cycle = 60%
 $T = T_{on} + T_{off} = 1 \text{ ms}$
 Duty Cycle = 0.6

Duty Cycle = $\frac{T_{on}}{T_{on} + T_{off}} = 0.6$
 $0.6 = \frac{T_{on}}{1 \text{ ms}}$
 $T_{on} = 0.6 \text{ ms}$
 $T_{off} = 1 \text{ ms} - 0.6 \text{ ms} = 0.4 \text{ ms}$
 Compute R_A and R_B using formula
 $T_{on} = 0.693 R_A C$
 $T_{off} = 0.693 R_B C$
 Choose $C = 0.1 \mu\text{F}$
 $R_A = 2.55 \text{ k}\Omega$ $R_B = 5.7 \text{ k}\Omega$

SYMMETRIC ASTABLE MULTIVIBRATOR:

Asymetric Frequency = 1 kHz and Duty Cycle = 50%
 $T = T_{on} + T_{off} = 1 \text{ ms}$
 Duty Cycle = 0.5

Duty Cycle = $\frac{T_{on}}{T_{on} + T_{off}} = 0.5$
 $0.5 = \frac{T_{on}}{1 \text{ ms}}$
 $T_{on} = 0.5 \text{ ms}$
 $T_{off} = 1 \text{ ms} - 0.5 \text{ ms} = 0.5 \text{ ms}$
 Compute T_{on} and T_{off} using formula
 $T_{on} = 0.693 R_A C$
 $T_{off} = 0.693 R_B C$
 $R_A = R_B = 7.2 \text{ k}\Omega$

ASTABLE MULTIVIBRATOR #AstableMultivibrator, #Multivibrator, #Timer

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- 3 FULL SUBTRACTOR #fullsubtractor #subtractor... SMVITM BANTAKAL 4:37
- 4 FULL SUBTRACTOR USING NAND... SMVITM BANTAKAL 3:40
- 5 MONOSTABLE MULTIVIBRATOR... SMVITM BANTAKAL 2:47
- 6 Narrow Band Reject Filter #Filters SMVITM BANTAKAL 13:47

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FULL ADDER USING NAND #NAND #fulladder #multisim

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ASTABLEMULTI - Multisim - (ASTABLEMULTI-Description)

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Design Toolbar

Design1
ASTABLEMULTI
ASTABLEMULTI-Description

ASYMMETRIC ASTABLE MULTIVIBRATOR

Assume Frequency = 1 kHz and Duty Cycle = 60%

$$T = T_{ON} + T_{OFF} = 1 \text{ ms}$$

Duty Cycle = 0.6

$$\text{Duty Cycle} = \frac{T_{ON}}{T_{ON} + T_{OFF}} = 0.6$$

so $T_{ON} = 0.6$ and $T_{OFF} = 0.4$

Compute R_A and R_B using formula
 $T_{ON} = 0.693(R_A + R_B)C$
 $T_{OFF} = 0.693R_B C$
 Choose $C = 0.1 \mu\text{F}$
 $R_A = 2.8\text{K}$ and $R_B = 5.7\text{K}$

SYMMETRIC ASTABLE MULTIVIBRATOR:

Assume Frequency = 1 kHz and Duty Cycle = 50%

$$T = T_{ON} + T_{OFF} = 1 \text{ ms}$$

Duty Cycle = 0.5

$$\text{Duty Cycle} = \frac{T_{ON}}{T_{ON} + T_{OFF}} = 0.5$$

Compute T_{ON} and T_{OFF} using formula
 $T_{ON} = 0.693R_A C$
 $T_{OFF} = 0.693R_B C$
 $R_A = R_B = 7.2\text{K}$

clideo.com

ASTABLEMULTI-Description

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Announce something to your class



Arun Upadhyaya ECE

Jul 4

Watch the video on the topic of Minimum Shift Keying for the next class

<https://youtu.be/2HVVPiKF0k8>

<https://youtu.be/EZPqLdt-Uk>

<https://youtu.be/ODo5wOB1sIM>



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SUMMARY ON NON IDEAL
TRANSISTOR.

The Shokely Transistor model derived a current-voltage expression for an ideal transistor. Some of the non ideal characteristics are:

- 1) Velocity Saturation: Electron velocity is related to electric field through mobility by the equation $v = \mu E$, where \bar{E} is the lateral electric field between drain & source. At higher E , μ is no more constant and it varies and is due to velocity saturation effect. As channel length becomes shorter, lateral electric field increases drain current and transistor becomes more velocity saturated and this increases and this decreases drain current. I_{ds}
- 2) Mobility degradation:- Velocity of charge carriers depend on electric field and when the carriers travel along the length of the channel, they get attracted to the surface by the vertical electric field. Therefore they bounce against the surface during their travel. This reduces surface mobility decreases in comparison. With the mobility along the channel as mobility decreases current also decreases.

- 3) Channel length modulation:- Ideally drain current I_{ds} is dependent on V_{ds} in the saturation region making transistor a perfect current source. when V_{ds} is increased further near the drain barrier is built due to depletion region & reduces the length of the channel. This result in reducing the length of the channel by ΔL . Thus in saturation saturation the effective channel length is $L_{eff} = L (1 - \frac{V_{ds}}{V_{dsat}})$.

- 4) Body effect:- MOSFET have 4th implicit terminal called body/substrate along with gate, source & drain. The



Threshold voltage V_t which is assumed to remain constant is no more a constant value & varies as potential b/w source and body is varied. Body effect parameter γ depends on doping level concentration. Thus by varying γ threshold voltage can be varied.

v) Sub Threshold conduction:- The ideal IV model ensures current flow from source to drain only when $v_{gs} > V_t$. But in practical transistor, current does not abruptly cut off below threshold but rather drops exponentially $v_{gs} < V_t$ is called weak inversion/ Subthreshold conduction is useful for designing low power analog circuits & dynamic circuits

vi) Drain Induced Barrier lowering (DIBL)

As the drain voltage v_{ds} is increased it creates an electric field that effects the threshold voltage is especially pronounced as in short channel transistors.

vii) Junction Leakage:- The mos structure is considered there exists pn junction b/w diffusion and the substrate. With cmos structures pn junction b/w diffusion and the substrate or well forming diodes. Leakage current usually lie in the range $0.1 - 0.01 \text{ fA}/\mu\text{m}^2$. which is negligibly when compared to other leakage currents

viii) Temperature Dependence:- With increase in temperature drain current decreases with temperature when transistor is ON. & when transistor is off, the junction leakage & subthreshold conduction contributes to leakage current and this increase.

ix) Geometry dependence:- The layout designer would draw transistor with width W & length L drawn W & L drawn while mask preparation the actual

length gate dimensions may differ by x_0 and x_L .
If the variations in the length & width of the transistor there will be variations in the performance.

Summary on Non-ideal Transistor I-V effects

The Shockley transistor model derived a current-voltage expression for an ideal transistor. Some of the non-ideal characteristics are:

- 1) mobility degradation: $v = \mu E$ where v is the carrier drift velocity and μ is mobility. The mobility term is independent of the vertical electric field E . The causes of mobility degradation is, when stronger vertical electric fields can attract charge carriers travelling along the inversion layer & cause it to collide with the oxide interface.
- 2) Velocity saturation: under strong lateral electric fields, the charge carriers are accelerated & the collisions between the charge carriers and silicon lattice increases. Beyond certain level of electric field, the carrier drift velocity saturates at v_{sat} . This is called velocity saturation.
- 3) Body effect: If the body is not connected to the source and it is connected to a different potential, then this causes the threshold voltage to vary.
- 4) Drain Induced barrier lowering (DIBL): The drain and the body form a PN junction, if the voltage on the drain is positive & the voltage on the body is negative, then this PN junction is reverse biased with increase in V_{ds} , reverse bias keeps increasing there by depletion region increases in size & start intruding on the channel.
- 5) Junction leakage: It refers to the undesired flow of current through the junctions of a semiconductor device, such as a diode or a transistor.
- 6) Short channel effects: It occurs in electronic devices due to the reduced channel length.

length in transistors. 7) Sub-threshold conduction: Also known as sub threshold leakage. It refers to the flow of current in a transistor, when it is operating below the threshold voltage, the transistor is not fully turned on, but there is still a small amount of leakage current passing through channel. 8) channel length modulation: As V_{ds} increases, the depletion region increases and the channel length becomes shorter. The current is inversely proportional to the length of the channel. 9) Gate leakage: The gate and the channel form a capacitor. Ideally no current should pass through the capacitor. It can be reduced by substituting the silicon dioxide with higher dielectric constant material.

Principals

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SUMMARY

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Laplace transform is a particular integral transform. This can be used for solving linear differential equations. Laplace transform can convert many common functions such as sinusoidal functions, damped sinusoidal functions, and exponential functions into algebraic functions of complex variable s .

Operations such as differentiation and integration can be replaced by algebraic operations in the complex plane. Thus, a linear differential equation can be transformed into an algebraic equation in complex variable s .

Applications of Laplace transform is to solve differential equations (both ordinary and partial), applications to RLC circuit analysis. Laplace transform converts differential equations in the time domain to algebraic equations in the frequency domain. (i) transformation of the time to frequency domain, (ii) manipulate the algebraic equations of the form a solution, (iii) inverse transformation from frequency to time domain.

Laplace transform formula is given by,

$$F(s) = \int_0^{\infty} f(t) e^{-st} dt$$

$f(s) \rightarrow$ Laplace transform

$s \rightarrow$ Complex number

$t \rightarrow$ real number ≥ 0

$dt \rightarrow$ first derivative of the function $f(t)$



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Summary

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Laplace transform is the integral transform of the given derivative function with real variable to convert into a complex function with variable s . It is useful in reducing the solution of a ordinary linear differential equation with constant coefficients to the solution of a polynomial equation. In Laplace transform, we calculate or analyse the circuit based on their frequency and time properties.

We have many number of properties in Laplace transform they are linear property, shifting property and so on. We are having inverse Laplace transform also the Laplace transform of $\sin at$ is $\frac{a}{s^2+a^2}$. The Laplace transform of $\cos at$ is $\frac{s}{s^2+a^2}$.



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Summary of Laplace Transform :-

The Laplace transform allows equations in the "time domain" to be transformed into an equivalent equation in the complex s Domain.

$\mathcal{L}\{f(t)\} = F(s)$, where the letter 's' has no special significance, and is used with the Laplace transform as a matter of common convention.

Laplace transform is the integral transform of the given derivative variable function with real variable t to convert into a complex function with variable s .

$$\mathcal{L}\{f(t)\} = \int_0^{\infty} f(t) e^{-st} dt.$$

The properties of Laplace transform are :-

* Linearity: The transform of a finite sum of time functions is the sum of Laplace transforms of the individual functions

$$\mathcal{L}\{f_1(t) + f_2(t) + f_3(t) + \dots + f_n(t)\} = F_1(s) + F_2(s) + F_3(s) + \dots + F_n(s)$$

* Scaling theorem: (Multiplication by K).

If K is a constant, then

$$\mathcal{L}\{K f(t)\} = K * \mathcal{L}\{f(t)\} = K * F(s).$$

* Real differentiation:

$$\mathcal{L}\left\{\frac{df(t)}{dt}\right\} = sF(s) - f(0^-) \text{ where } f(0^-) \text{ is the value of } f(t) \text{ when } t=0^-.$$

* Real Integration property:

$$\mathcal{L}\left\{\int_0^t f(t) dt\right\} = \frac{F(s)}{s}$$

* Differentiation by 's' (Multiplication by 't') property.

$$\mathcal{L}\{t f(t)\} = -\frac{dF(s)}{ds}$$

* Complex Translation property.

$\mathcal{L}\{e^{at} f(t)\} = F(s-a)$ where 'a' is a complex number &

$$\mathcal{L}\{e^{-at} f(t)\} = F(s+a).$$

* Real Translation (Shifting) theorem:

$$\mathcal{L}\{f(t-T)\} = e^{-Ts} F(s).$$

* Initial Value theorem:

$$f(0^+) = \lim_{t \rightarrow 0^+} f(t) = \lim_{s \rightarrow \infty} sF(s)$$

* Final Value theorem:

$$\lim_{t \rightarrow \infty} f(t) = \lim_{s \rightarrow 0} sF(s).$$

* Convolution theorem:

$$f_1(t) * f_2(t) = \int_0^t f_1(\tau) f_2(t-\tau) d\tau = \int_0^t f_2(\tau) f_1(t-\tau) d\tau$$

↑
convolution

where τ is the dummy variable.

$$\therefore \mathcal{L}\{f_1(t) * f_2(t)\} = F_1(s) \cdot F_2(s).$$

There are four basic Standard Time functions :-

* Step $\Rightarrow \mathcal{L}\{u(t)\} = 1/s$

* Ramp $\Rightarrow \mathcal{L}\{r(t)\} = 1/s^2$

* Parabolic ~~$\Rightarrow \mathcal{L}\{s(t)\} = 1/s^3$~~

* Impulse $\Rightarrow \mathcal{L}\{\delta(t)\} = 1$

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* Step function is a function which has a value of A for $t \geq 0$ and 0 for $t < 0$.

* Ramp function is one which has slope ' A ' where ' A ' is called magnitude or amplitude of ramp function.

* Impulse function is one which exists only at $t=0$ and is zero elsewhere. \Rightarrow It is also called Delta function $\{\delta(t)\}$.

Then,

$r(t)$ Differentiate \rightarrow $u(t)$ Differentiate \rightarrow $\delta(t)$.
 ramp function unit step function Impulse function

$\delta(t)$ Integrate \rightarrow $u(t)$ Integrate \rightarrow $r(t)$.

Laplace transform of periodic function,

$$F(s) = \frac{1}{1 - e^{-sT}} \cdot F_1(s).$$

Inverse Laplace Transform:-

If $L\{f(t)\} = F(s)$, then

$$L^{-1}\{F(s)\} = f(t).$$

Two cases:

i) If Degree of $N(s) <$ degree of $D(s)$, then apply partial fraction method directly.

$$F(s) = \frac{N(s)}{D(s)}$$

ii) If degree of $N(s) \geq$ degree of $D(s)$, then, divide $N(s)$ by $D(s)$, then apply partial fraction method to second term.

$$F(s) = Q + F_1(s) = Q + \frac{N'(s)}{D'(s)}.$$


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Application of Laplace transform:

Laplace transform can be used to transform the time domain circuits into s domain circuits to simplify the solution of integral differential equations to the manipulation of a set of algebraic equations.

* * *


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